HEF4046B

Phase-locked loop

Rev. 8 — 3 September 2024

Product data sheet

1. General description

The HEF4046B is a phase-locked loop circuit that consists of a linear voltage controlled oscillator (VCO) and two different phase comparators with a common signal input amplifier and a common comparator input.

2. Features and benefits

- Wide supply voltage range from 3.0 V to 15.0 V
- CMOS low power dissipation
- · High noise immunity
- · Fully static operation
- 5 V, 10 V, and 15 V parametric ratings
- · Standardized symmetrical output characteristics
- · Complies with JEDEC standard JESD 13-B
- ESD protection:
 - HBM: ANSI/ESDA/JEDEC JS-001 class 2 exceeds 2000 V
 - CDM: ANSI/ESDA/JEDEC JS-002 class C3 exceeds 1000 V
- Specified from -40 °C to +85 °C

3. Ordering information

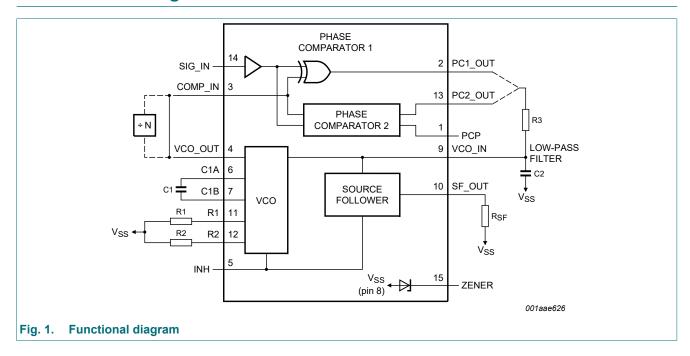
Table 1. Ordering information

Type number	Package						
	Temperature range	Name	Description	Version			
HEF4046BT	-40 °C to +85 °C	SO16	plastic small outline package; 16 leads; body width 3.9 mm	SOT109-1			



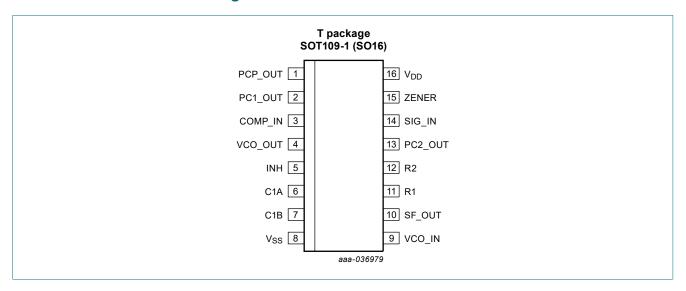
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4. Functional diagram



5. Pinning information

5.1. Pinning



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5.2. Pin description

Table 2. Pin description

Symbol	Pin	Description		
PCP_OUT	1	phase comparator pulse output		
PC1_OUT	2	phase comparator 1 output		
COMP_IN	3	comparator input		
VCO_OUT	4	VCO output		
INH	5	inhibit input		
C1A	6	capacitor C1 connection A		
C1B	7	capacitor C1 connection B		
V _{SS}	8	ground supply voltage		
VCO_IN	9	VCO input		
SF_OUT	10	source-follower output		
R1	11	resistor R1 connection		
R2	12	resistor R2 connection		
PC2_OUT	13	phase comparator 2 output		
SIG_IN	14	signal input		
ZENER	15	Zener diode input for regulated supply		
V_{DD}	16	supply voltage		

6. Functional description

6.1. VCO control

The VCO requires an external capacitor (C1) and resistor (R1) with an optional resistor (R2). Resistor R1 and capacitor C1 determine the frequency range of the VCO, while resistor R2 enables the VCO to have a frequency off-set if required. The high input impedance of the VCO simplifies the design of low-pass filters; it permits the designer a wide choice of resistor/capacitor ranges. In order not to load the low-pass filter, a source-follower output of the VCO input voltage is provided at SF_OUT (pin 10). If this is used, a load resistor (R_L) should be connected from SF_OUT to V_{SS}; if unused, SF_OUT should be left open. The VCO output (pin 4) can either be connected directly to the comparator input COMP_IN (pin 3) or via a frequency divider. A LOW-level at the inhibit input INH_IN (pin 5) enables the VCO and the source follower, while a HIGH-level turns both off to minimize standby power consumption.

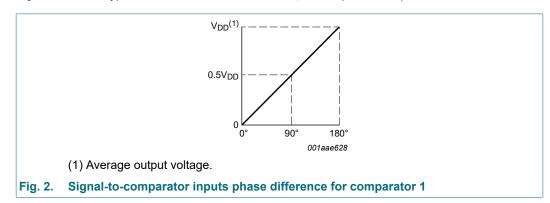
6.2. Phase comparators

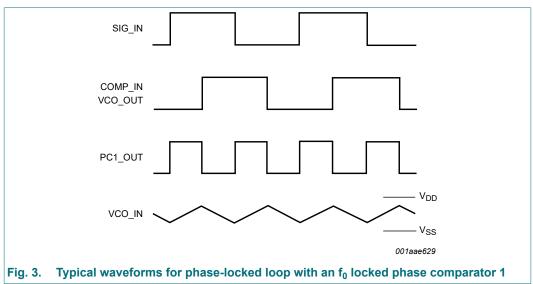
The phase-comparator signal input SIG_IN (pin 14) can be direct-coupled, provided the signal swing is between the standard HEF4000B family input logic levels. The signal must be capacitively coupled to the self-biasing amplifier at the signal input with smaller swings. Phase comparator 1 is an EXCLUSIVE-OR network. The signal and comparator input frequencies must have a 50% duty factor to obtain the maximum lock range. The average output voltage of the phase comparator is equal to $0.5V_{DD}$ when there is no signal or noise at the signal input. The average voltage to the VCO input VCO_IN is supplied by the low-pass filter connected to the output of phase comparator 1. This also causes the VCO to oscillate at the center frequency (f_0). The frequency capture range (f_0) is defined as the frequency range of input signals on which the PLL will lock if it was initially out of lock. The frequency lock range (f_0) is defined as the frequency range of input signals on which the loop will stay locked if it was initially in lock. The capture range is smaller or equal to the lock range.

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With phase comparator 1, the range of frequencies over which the PLL can acquire lock (capture range) depends on the low-pass filter characteristics and this range can be made as large as the lock range. Phase comparator 1 enables the PLL system to remain in lock in spite of high amounts of noise in the input signal. A typical behavior of this type of phase comparator is that it may lock onto input frequencies that are close to harmonics of the VCO center frequency. Another typical behavior is that the phase angle between the signal and comparator input varies between 0° and 180°, and is 90° at the center frequency. Fig. 2 shows the typical phase-to-output response characteristic.

Fig. 3 shows the typical waveforms for a PLL with a f₀ locked phase comparator 1.



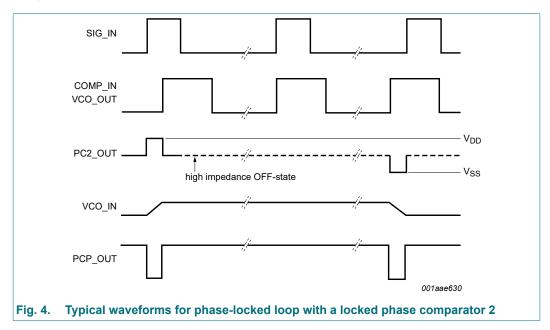


Phase comparator 2 is an edge-controlled digital memory network. It consists of four flip-flops, control gating and a 3-state output circuit comprising p and n-type drivers with a common output node. When the p-type or n-type drivers are ON, they pull the output up to V_{DD} or down to V_{SS} respectively. This type of phase comparator only acts on the positive-going edges of the signals at SIG_IN and COMP_IN. Therefore, the duty factors of these signals are not of importance.

If the signal input frequency is higher than the comparator input frequency, the p-type output driver is maintained ON most of the time, and both the n and p-type drivers are OFF (3-state) the remainder of the time. If the signal input frequency is lower than the comparator input frequency, the n-type output driver is maintained ON most of the time, and both the n and p-type drivers are OFF the remainder of the time. If the signal input and comparator input frequencies are equal, but the signal input lags the comparator input in phase, the n-type output driver is maintained ON for a time corresponding to the phase difference. If the comparator input lags the signal input in phase, the p-type output driver is maintained ON for a time corresponding to the phase difference. Subsequently, the voltage at the capacitor of the low-pass filter connected to this phase comparator is adjusted until the signal and comparator inputs are equal in both phase and frequency. At this stable point, both p and n-type drivers remain OFF and thus the phase comparator output becomes an open circuit and keeps the voltage at the capacitor of the low-pass filter constant.

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Moreover, the signal at the phase comparator pulse output (PCP_OUT) is a HIGH level, which can be used for indicating a locked condition. Thus, for phase comparator 2, no phase difference exists between the signal and comparator inputs over the full VCO frequency range. Moreover, the power dissipation due to the low-pass filter is reduced when this type of phase comparator is used, because both p and n-type output drivers are OFF for most of the signal input cycle. It should be noted that the PLL lock range for this type of phase comparator is equal to the capture range, independent of the low-pass filter. With no signal present at the signal input, the VCO is adjusted to its lowest frequency for phase comparator 2. Fig. 4 shows typical waveforms for a PLL employing this type of locked phase comparator.



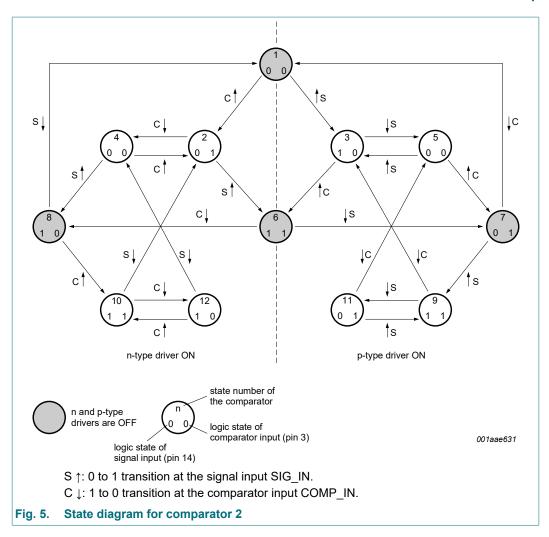
<u>Fig. 5</u> shows the state diagram for phase comparator 2. Each circle represents a state of the comparator. The number at the top, inside each circle, represents the state of the comparator, while the logic state of the signal and comparator inputs are represented by a '0' for a logic LOW or a '1' for a logic HIGH, and they are shown in the left and right bottom of each circle.

The transitions from one to another result from either a logic change at the signal input (S representing SIG_IN) or the comparator input (C representing COMP_IN). A positive- going and a negative-going transition are shown by an arrow pointing up or down respectively.

The state diagram assumes, that only one transition on either the signal input or comparator input occurs at any instant.

- States 3, 5, 9 and 11 represent the output condition when the p-type driver is ON.
- States 2, 4, 10 and 12 determine the condition when the n-type driver is ON.
- States 1, 6, 7 and 8 represent the condition when the output is in its high-impedance OFF state;
 i.e. both p and n-type drivers are OFF, and the PCP_OUT output is HIGH. The condition at output PCP_OUT for all other states is LOW.

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7. Limiting values

Table 3. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
V_{DD}	supply voltage		-0.5	+18	V
I _{IK}	input clamping current	$V_{I} < -0.5 \text{ V or } V_{I} > V_{DD} + 0.5 \text{ V}$	-	±10	mA
V_{I}	input voltage		-0.5	V _{DD} + 0.5	V
I _{OK}	output clamping current	$V_O < -0.5 \text{ V or } V_O > V_{DD} + 0.5 \text{ V}$	-	±10	mA
I _{I/O}	input/output current		-	±10	mA
I _{DD}	supply current		-	50	mA
T _{stg}	storage temperature		-65	+150	°C
T _{amb}	ambient temperature		-40	+85	°C
P _{tot}	total power dissipation	-40 °C to +85 °C	-	500	mW
Р	power dissipation	per output	-	100	mW

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8. Recommended operating conditions

Table 4. Recommended operating conditions

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V_{DD}	supply voltage		3	-	15	V
		as fixed oscillator only	3	-	15	V
		phase-locked loop operation	5	-	15	V
VI	input voltage		0	-	V_{DD}	V
T _{amb}	ambient temperature	in free air	-40	-	+85	°C
Δt/ΔV	input transition rise and fall rate	for INH input				
		V _{DD} = 5 V	-	-	3.75	µs/V
		V _{DD} = 10 V	-	-	0.5	μs/V
		V _{DD} = 15 V	-	-	0.08	μs/V

9. Static characteristics

Table 5. Static characteristics

 $V_{SS} = 0 \ V$; $V_I = V_{SS}$ or V_{DD} unless otherwise specified.

Symbol	Parameter	neter Conditions \	V _{DD}	T _{amb} =	-40 °C	T _{amb} =	+25 °C	T _{amb} = +85 °C		Unit
				Min	Max	Min	Max	Min	Max	
V _{IH}	HIGH-level	I _O < 1 μΑ	5 V	3.5	-	3.5	-	3.5	-	V
	input voltage	out voltage	10 V	7.0	-	7.0	-	7.0	-	V
			15 V	11.0	-	11.0	-	11.0	-	V
V _{IL}	LOW-level	I _O < 1 μΑ	5 V	-	1.5	-	1.5	-	1.5	V
	input voltage		10 V	-	3.0	-	3.0	-	3.0	V
			15 V	-	4.0	-	4.0	-	4.0	V
V _{OH}	HIGH-level	I _O < 1 μΑ	5 V	4.95	-	4.95	-	4.95	-	V
	output voltage		10 V	9.95	-	9.95	-	9.95	-	V
			15 V	14.95	-	14.95	-	14.95	-	V
V _{OL}	LOW-level output voltage	I _O < 1 μΑ	5 V	-	0.05	-	0.05	-	0.05	V
		output voltage	10 V	-	0.05	-	0.05	-	0.05	V
			15 V	-	0.05	-	0.05	-	0.05	V
I _{OH}	HIGH-level	V _O = 2.5 V	5 V	-	-1.7	-	-1.4	-	-1.1	mA
	output current	V _O = 4.6 V	5 V	-	-0.52	-	-0.44	-	-0.36	mA
		V _O = 9.5 V	10 V	-	-1.3	-	-1.1	-	-0.9	mA
		V _O = 13.5 V	15 V	-	-3.6	-	-3.0	-	-2.4	mA
I _{OL}	LOW-level output	V _O = 0.4 V	5 V	0.52	-	0.44	-	0.36	-	mA
	current	V _O = 0.5 V	10 V	1.3	-	1.1	-	0.9	-	mA
		V _O = 1.5 V	15 V	3.6	-	3.0	-	2.4	-	mA
I _I	input leakage current		15 V	-	±0.3	-	±0.3	-	±1.0	μΑ
I _{OZ}	OFF-state output current	output HIGH and returned to V _{DD}	15 V	-	1.6	-	1.6	-	12.0	μA
		output LOW and returned to V _{SS}	15 V	-	1.6	-	1.6	-	12.0	μΑ

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Symbol	Parameter	Conditions	V_{DD}		T _{amb} =	-40 °C	T _{amb} =	+25 °C	T _{amb} =	+85 °C	Unit
					Min	Max	Min	Max	Min	Max	
I_{DD}	supply current	I _O = 0 A	5 V	[1]	-	-	20	-	-	-	μΑ
			10 V	[1]	-	-	300	-	-	-	μΑ
			15 V	[1]	-	-	750	-	-	-	μΑ
			5 V	[2]	-	20	-	20	-	150	μΑ
			10 V	[2]	-	40	-	40	-	300	μΑ
			15 V	[2]	-	80	-	80	-	600	μΑ
Cı	input capacitance	for INH input			-	-	-	7.5	-	-	pF

- Pin 15 open; pin 5 at V_{DD} ; pins 3 and 9 at V_{SS} ; pin 14 open. Pin 15 open; pin 5 at V_{DD} ; pins 3 and 9 at V_{SS} ; pin 14 at V_{DD} ; input current at pin 14 not included.

10. Dynamic characteristics

Table 6. Dynamic characteristics

 V_{SS} = 0 V; T_{amb} = 25 °C; C_L = 50 pF; input transition times \leq 20 ns.

Symbol	Parameter	Conditions	V _{DD}	Min	Тур	Max	Unit	
Phase comparators								
R _I	input resistance	out resistance SIG_IN input; at self-bias operating point	5 V	-	750	-	kΩ	
			10 V	-	220	-	kΩ	
			15 V	-	140	-	kΩ	
V _{i(sens)}	input voltage	SIG_IN input, AC coupled, peak-to-peak	5 V	-	150	-	mV	
	sensitivity	values; R1 = 10 kΩ; R2 = ∞ ; C1 = 100 pF; independent of the lock range	10 V	-	150	-	mV	
		indopondent of the look range	15 V	-	200	-	mV	
V _{IL}	LOW-level input	SIG_IN and COMP_IN inputs, DC	5 V	-	-	1.5	V	
	voltage	coupled LOW; full temperature range	10 V	-	-	3.0	V	
			15 V	-	-	4.0	V	
V _{IH}	HIGH-level input	SIG_IN and COMP_IN inputs, DC	5 V	3.5	-	-	V	
	voltage	coupled HIGH; full temperature range	10 V	7.0	-	-	V	
			15 V	11.0	-	-	V	
I _{IH}	HIGH-level input	SIG_IN input; at V _{DD}	5 V	-	7	-	μA	
	current		10 V	-	30	-	μA	
				-	70	-	μA	
I _{IL}	LOW-level input SIG_IN input; at V _{SS}		5 V	-	-3	-	μA	
	current		10 V	-	-18	-	μA	
			15 V	-	-45	-	μA	

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Symbol	Parameter	Conditions	V_{DD}	Min	Тур	Max	Unit
vco							
Р	power dissipation $f_0 = 10 \text{ kHz}$; R1 = 1 M Ω ; R2 = ∞ ; VCO IN at 0.5 Vpp;		5 V	-	150	-	μW
		VCO_IN at 0.5 V _{DD} ; see <u>Fig. 9</u> , <u>Fig. 10</u> , and <u>Fig. 11</u>	10 V	-	2500	-	μW
		See <u>Fig. 9</u> , <u>Fig. 10</u> , and <u>Fig. 11</u>	15 V	-	9000	-	μW
f _{max}	maximum frequency	VCO_IN at V _{DD} ; R1 = 10 kΩ; R2 = ∞;	5 V	0.5	1.0	-	MHz
		C1 = 50 pF	10 V	1.0	2.0	-	MHz
			15 V	1.3	2.7	-	MHz
Δf/ΔΤ	frequency variation with temperature	no frequency offset (f _{min} = 0 Hz)	5 V [1]	-	0.22 to 0.30	-	% Hz/°C
			10 V [1]	-	0.04 to 0.05	-	% Hz/°C
			15 V [1]	-	0.01 to 0.05	-	% Hz/°C
		with frequency offset (f _{min} > 0 Hz)	5 V [1]	-	0 to 0.22	-	% Hz/°C
			10 V [1]	-	0 to 0.04	-	% Hz/°C
			15 V [1]	-	0 to 0.01	-	% Hz/°C
Δf/f	relative frequency	for VCO see Fig. 12 and Fig. 13					
	variation	R1 > 10 kΩ	5 V	-	0.50	-	% Hz
		R1 > 400 kΩ	10 V	-	0.25	-	% Hz
		R1 = 1 MΩ	15 V	-	0.25	-	% Hz
δ	duty factor	VCO _OUT output	5 V	-	50	-	%
				-	50	-	%
			15 V	-	50	-	%
R _{in}	input resistance	for pin VCO_IN			10		МΩ
Source f	ollower						
V_{offset}	offset voltage	$R_L = 10 \text{ k}\Omega; VCO_IN \text{ at } 0.5V_{DD}$	5 V [2]	-	1.7	-	V
			10 V	-	2.0	-	V
			15 V	-	2.1	-	V
		$R_L = 50 \text{ k}\Omega; VCO_IN \text{ at } 0.5V_{DD}$	5 V	-	1.5	-	V
			10 V	-	1.7	-	V
			15 V	-	1.8	-	V
∆f/f	relative frequency	VCO output; $R_L > 50 \text{ k}\Omega$; see Fig. 12	5 V 10 V	-	0.3	-	%
	variation	riation		-	1.0	-	%
			15 V	-	1.3	-	%
Zener di							
V_Z	working voltage	I _Z = 50 μA	-	-	7.3	-	V
R _{dyn}	dynamic resistance	For internal zener diode; I _Z = 1 mA	-	-	25	-	Ω

^[1] Over the recommended component range.

The offset voltage is equal to the input voltage on pin VCO_IN minus the output voltage on pin SF_OUT.

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11. Design information

Table 7. Design information

Test	Using phase comparator 1	Using phase comparator 2	
VCO adjusts with no signal on SIG_IN	VCO in PLL system adjusts to center frequency (f ₀)	VCO in PLL system adjusts to minimum frequency (f _{min})	
Phase angle between SIG_IN and COMP_IN	90° at center frequency (f ₀), approaching 0° and 180° at the ends of the lock range (2f _L)	always 0° in lock (positive-going edges)	
Locks on harmonics of center frequency	yes	no	
Signal input noise rejection	high	low	
Lock frequency range (2f _L)	the frequency range of the input signa it was initially in lock; 2f _L = full VCO fre		
Capture frequency range (2f _c)	the frequency range of the input signa initially out of lock	l on which the loop will lock if it was	
	depends on low-pass filter characteristics; 2f _c < 2f _L	$2f_c = 2f_L$	
Center frequency (f ₀)	the frequency of the VCO when VCO_IN at 0.5V _{DD}		

11.1. VCO component selection

Recommended range for R1 and R2: 10 k Ω to 1 M Ω .

Recommended range for C1: 50 pF to any practical value.

- 1. VCO without frequency offset (R2 = ∞).
 - **a.** Given f_0 : use f_0 with Fig. 6 to determine R1 and C1.
 - **b.** Given f_{max} : calculate f_0 from $f_0 = 0.5 f_{max}$; use f_0 with Fig. 6 to determine R1 and C1.
- 2. VCO with frequency offset.
 - **a.** Given f_0 and $2f_L$: calculate f_{min} from the equation f_{min} = f_0 $2f_L$; use f_{min} with Fig. 7 to determine R2 and C1;

calculate
$$\frac{f_{\text{max}}}{f_{\text{min}}}$$
 from the equation $\frac{f_{\text{max}}}{f_{\text{min}}} = \frac{f_0 + 2f_L}{f_0 - 2f_L}$;

use $\frac{f_{\text{max}}}{f_{\text{min}}}$ with Fig. 8 to determine the ratio R2/R1 to obtain R1.

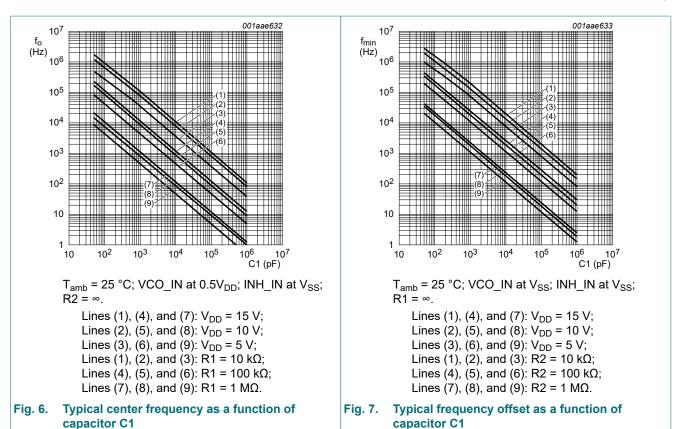
b. Given f_{min} and f_{max} : use f_{min} with Fig. 7 to determine R2 and C1;

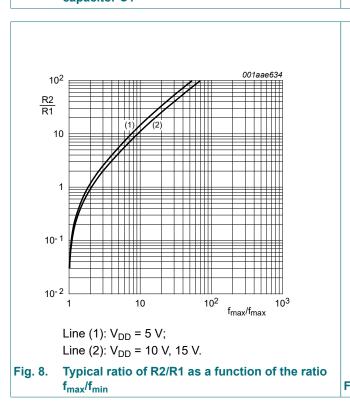
calculate
$$\frac{f_{\text{max}}}{f_{\text{min}}}$$
;

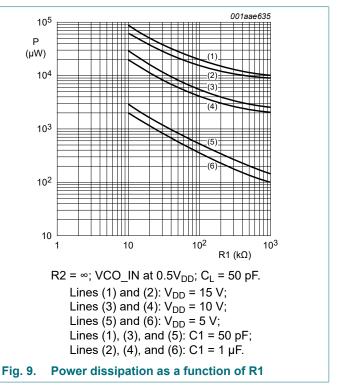
use
$$\frac{f_{\text{max}}}{f_{\text{min}}}$$
 with Fig. 8 to determine R2/R1 to obtain R1.

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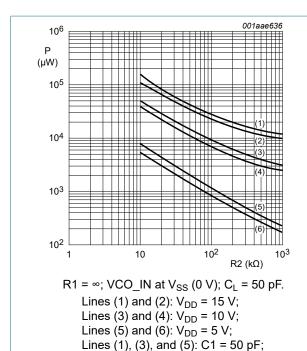


Fig. 10. Power dissipation as a function of R2

Lines (2), (4), and (6): C1 = 1 μ F.

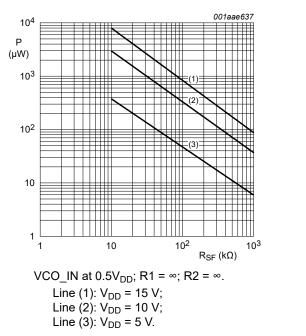
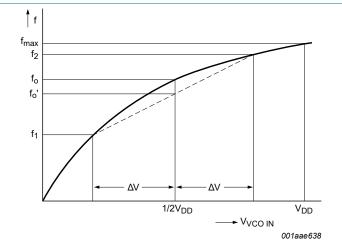


Fig. 11. Power dissipation of source follower as a function of \mathbf{R}_{L}



See Section 10.

For VCO linearity:

$$\mathbf{f'}_0 = \frac{f_1 + f_2}{2}$$

linearity = $\frac{f'_0 - f_0}{f'_0} \times 100 \%$

This figure and the above formula also apply to source follower linearity: substitute V_O at SF_OUT for f.

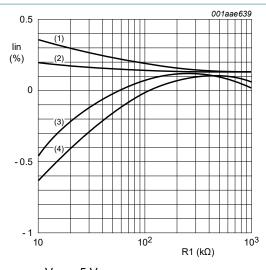
 ΔV = 0.3 V at V_{DD} = 5 V;

 ΔV = 2.5 V at V_{DD} = 10 V;

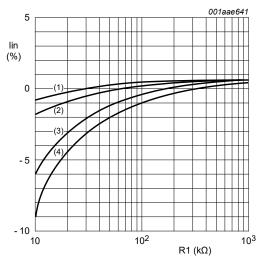
 $\Delta V = 5.0 \text{ V}$ at $V_{DD} = 15 \text{ V}$.

Fig. 12. Definition of linearity

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c. V_{DD} = 15 V

R2 = ∞;

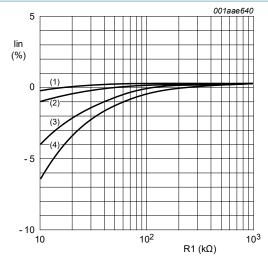
Line (1): $C1 = 1 \mu F$;

Line (2): C1 = 1 nF;

Line (3): C1 = 100 pF;

Line (4): C1 = 50 pF.

Fig. 13. VCO frequency linearity as a function of R1



b. V_{DD} = 10 V

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12. Package outline

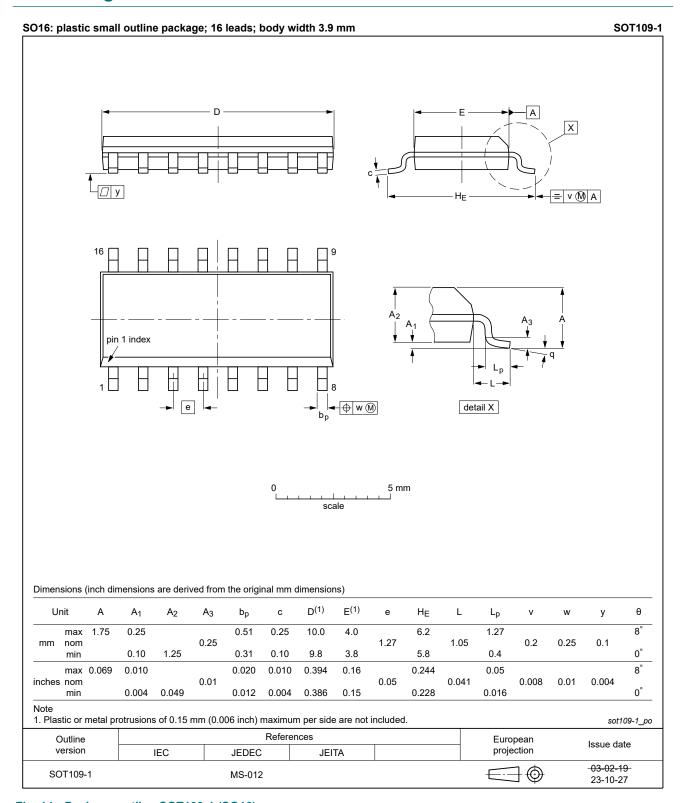


Fig. 14. Package outline SOT109-1 (SO16)

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13. Abbreviations

Table 8. Abbreviations

Acronym	Description	
ANSI	American National Standards Institute	
CDM	Charged Device Model	
CMOS	Complementary Metal-Oxide Semiconductor	
DUT	Device Under Test	
ESD	ElectroStatic Discharge	
ESDA	ElectroStatic Discharge Association	
НВМ	Human Body Model	
JEDEC	oint Electron Device Engineering Council	
PLL	Phase-Locked Loop	

14. Revision history

Table 9. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes				
HEF4046B v.8	20240903	Product data sheet	-	HEF4046B v.7.1				
Modifications:		 <u>Section 2</u>: ESD specification updated according to the latest JEDEC standard. <u>Fig. 14</u>: Aligned SO package outline drawing to JEDEC MS-012 						
HEF4046B v.7.1	20231020	Product data sheet	-	HEF4046B v.6				
Modifications:	guidelines of Legal texts Section 1 a	of this data sheet has been of Nexperia. have been adapted to the nd Section 2 updated. Section 11.1: errata. added.	•					
HEF4046B v.6	20160324	Product data sheet	-	HEF4046B v.5				
Modifications:	Type numb	er HEF4046BP (SOT38-4) removed.					
HEF4046B v.5	20111118	Product data sheet	-	HEF4046B v.4				
Modifications:	• <u>Table 5</u> : I _{OF}	 Section Applications removed Table 5: I_{OH} minimum values changed to maximum Table 6: R_{in} typical value changed from 10⁶ MΩ to 10 MΩ 						
HEF4046B v.4	20100105	Product data sheet	-	HEF4046B_CNV v.3				
HEF4046B_CNV v.3	19950101	Product specification	-	HEF4046B_CNV v.2				
HEF4046B_CNV v.2	19950101	Product specification	-	-				

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15. Legal information

Data sheet status

Document status [1][2]	Product status [3]	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

- Please consult the most recently issued document before initiating or completing a design.
- [2] The term 'short data sheet' is explained in section "Definitions".
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